

# **TDA7388**

# 4 X 41W QUAD BRIDGE CAR RADIO AMPLIFIER

# **1 FEATURES**

- HIGH OUTPUT POWER CAPABILITY:
- 4 x 41W/4Ω MAX.
- 4 x 25W/4Ω @ 14.4V, 1KHz, 10%
- LOW DISTORTION
- LOW OUTPUT NOISE
- ST-BY FUNCTION
- MUTE FUNCTION
- AUTOMUTE AT MIN. SUPPLY VOLTAGE DETECTION
- LOW EXTERNAL COMPONENT COUNT:
  - INTERNALLY FIXED GAIN (26dB)
  - NO EXTERNAL COMPENSATION
  - NO BOOTSTRAP CAPACITORS

# 2 PROTECTIONS:

- OUTPUT SHORT CIRCUIT TO GND, TO V<sub>S</sub>, ACROSS THE LOAD
- VERY INDUCTIVE LOADS
- OVERRATING CHIP TEMPERATURE WITH SOFT THERMAL LIMITER
- LOAD DUMP VOLTAGE
- FORTUITOUS OPEN GND

#### Figure 2. Block and Application Diagram

### Figure 1. Package



#### Table 1. Order Codes

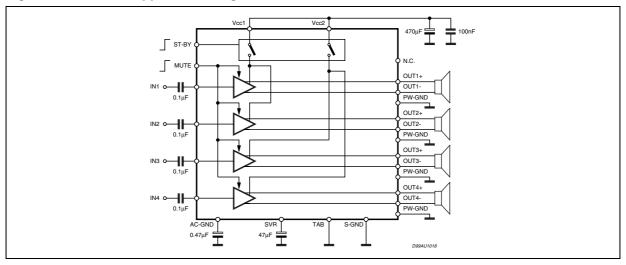
Part Number	Package
TDA7388	Flexiwatt25

- REVERSED BATTERY
- ESD

## **3 DESCRIPTION**

The TDA7388 is a new technology class AB Audio Power Amplifier in Flexiwatt 25 package designed for high end car radio applications.

Thanks to the fully complementary PNP/NPN output configuration the TDA7388 allows a rail to rail output voltage swing with no need of bootstrap capacitors. The extremely reduced components count allows very compact sets.

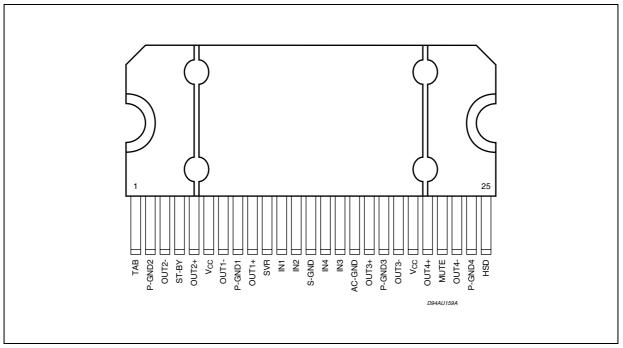


## TDA7388

Symbol	Parameter	Value	Unit
V <sub>CC</sub>	Operating Supply Voltage	18	V
V <sub>CC (DC)</sub>	DC Supply Voltage	28	V
V <sub>CC (pk)</sub>	Peak Supply Voltage (t = 50ms)	50	
Ι <sub>Ο</sub>	Output Peak Current: Repetitive (Duty Cycle 10% at f = 10Hz) Non Repetitive (t = $100\mu s$ )	4.5 5.5	A A
P <sub>tot</sub>	Power dissipation, (T <sub>case</sub> = 70°C)	80	W
Tj	Junction Temperature	150	°C
T <sub>stg</sub> Storage Temperature		– 55 to 150	°C

### **Table 2. Absolute Maximum Ratings**

# Figure 3. Pin Connection



## Table 3. Thermal Data

Syr	mbol	Parameter	Value	Unit
R <sub>th</sub>	n j-amb	Thermal Resistance Junction to Case max	1	°C/W

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
l <sub>q1</sub>	Quiescent Current	$R_L = \infty$	120	190	350	mA
V <sub>OS</sub>	Output Offset Voltage	Play Mode			±80	mV
$\Delta V_{OS}$	During Mute ON/OFF Output Offset Voltage				±80	mV
Gv	Voltage Gain		25	26	27	dB
Po	Output Power	THD = 10%; $V_S = 14.4V$	22	26		W
P <sub>o max</sub>	Max.Output Power (*)	V <sub>S</sub> = 14.4V	38	41		W
THD	Distortion	$P_0 = 4W$		0.04	0.15	%
e <sub>No</sub>	Output Noise	"A" Weighted		50	70	μV
		Bw = 20Hz to 20KHz		70	100	μV
SVR	Supply Voltage Rejection	$f = 100Hz; V_r = 1V_{rms}$	50	65		dB
f <sub>ch</sub>	High Cut-Off Frequency	$P_0 = 0.5W$	100	200		KHz
Ri	Input Impedance		70	100		KΩ
CT	Cross Talk	f = 1KHz; Po = 4W	60	70		dB
		f = 10KHz; Po = 4W	50	60		dB
I <sub>SB</sub>	St-By Current Consumption				50	μA
$V_{SB out}$	St-By OUT Threshold Voltage	(Amp: ON)	3.5			V
V <sub>SB IN</sub>	St-By IN Threshold Voltage	(Amp: OFF)			1.5	V
A <sub>M</sub>	Mute Attenuation	P <sub>Oref</sub> = 4W	80	90		dB
V <sub>M out</sub>	Mute OUT Threshold Voltage	(Amp: Play)	3.5			V
V <sub>M in</sub> Mute IN Threshold Voltage		(Amp: Mute)			1.5	V
V <sub>AM in</sub> V <sub>S</sub> Automute Threshold		$\begin{array}{l} \mbox{(Amp: Mute); Att } \geq 80dB; \ \mbox{P}_{Oref} = 4\Omega \\ \mbox{(Amp: Play); Att } < 0.1dB; \ \mbox{P}_{O} = 0.5\Omega \end{array}$		7.6	6.5 8.5	V V
I <sub>pin22</sub>	Muting Pin Current	V <sub>MUTE</sub> = 1.5V (Source Current)	5	11	20	μΑ

**Table 4. Electrical Characteristcs** ( $V_S = 14.4V$ ; f = 1KHz;  $R_g = 600\Omega$ ;  $R_L = 4\Omega$ ;  $T_{amb} = 25^{\circ}C$ ; Refer to the Test and application diagram, unless otherwise specified.)

(\*) Saturated square wave output.

## TDA7388

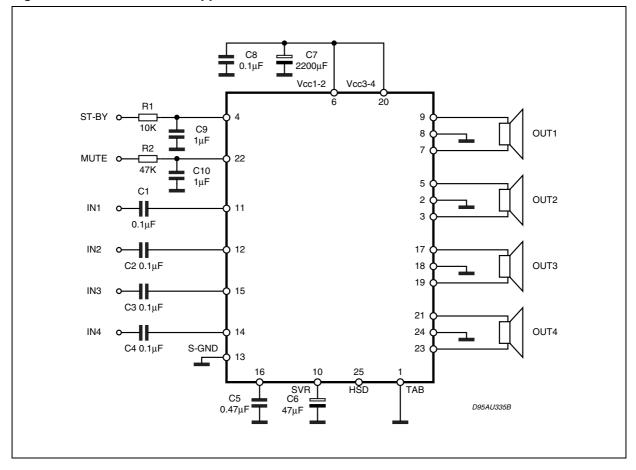


Figure 4. Standard Test and Application Circuit

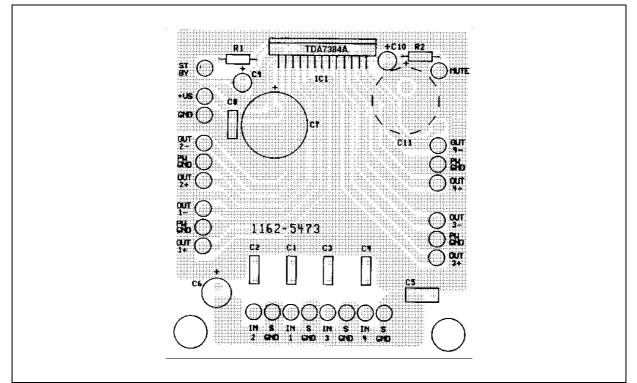


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# 4 P.C.B. AND COMPONENT LAYOUT OF THE FIGURE 4

## Figure 5. Components & Top Copper Layer



## Figure 6. Bottom Copper Layer

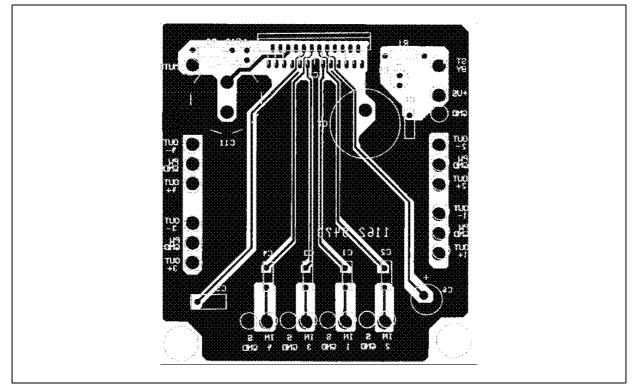
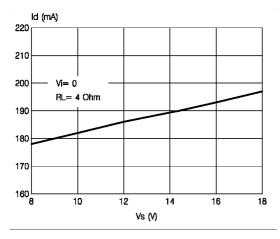
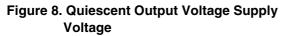
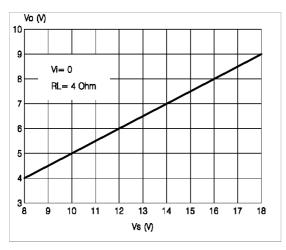


Figure 7. Quiescent Current vs. Supply Voltage









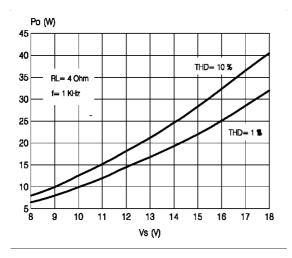


Figure 10. Distortion vs. Output Power

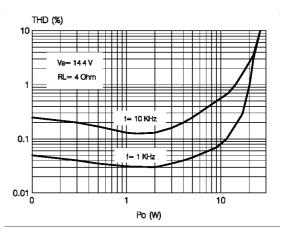
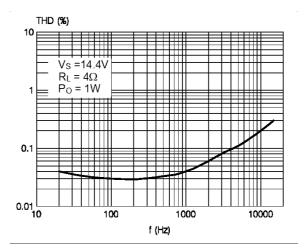
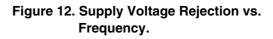
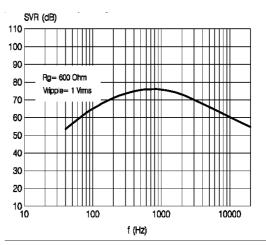


Figure 11. Distortion vs. Frequency







# Figure 13. Output Noise vs. Source Resistance.

 $\top$ 

1000

10000

100000

"A" wgtd

Rg (ohm)

100

22-22K Hz In

En (uV)

Vs= 14.4 V

PI = 4 Obr

120

110

100

90

80

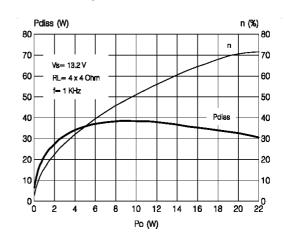
70

60

50

40

30 20



#### Figure 14. Power Dissipation & Efficiency vs. Output Power.

# **5 APPLICATION HINTS**

(ref. to the circuit of fig. 4)

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## 5.1 SVR

Besides its contribution to the ripple rejection, the SVR capacitor governs the turn ON/OFF time sequence and, consequently, plays an essential role in the pop optimization during ON/OFF transients. To conveniently serve both needs, **ITS MINIMUM RECOMMENDED VALUE IS 10\muF**.

### 5.2 INPUT STAGE

The TDA7388'S inputs are ground-compatible and can stand very high input signals ( $\pm$  8Vpk) without any performances degradation. If the standard value for the input capacitors (0.1µF) is adopted, the low frequency cut-off will amount to 16 Hz.

## 5.3 STAND-BY AND MUTING

STAND-BY and MUTING facilities are both CMOS-COMPATIBLE. If unused, a straight connection to Vs of their respective pins would be admissible.

Conventional/low-power transistors can be employed to drive muting and stand-by pins in absence of true CMOS ports or microprocessors. R-C cells have always to be used in order to smooth down the transitions for preventing any audible transient noises.

Since a DC current of about 10  $\mu$ A normally flows out of pin 22, the maximum allowable muting-series resistance (R2) is 70K $\Omega$ , which is sufficiently high to permit a muting capacitor reasonably small (about 1 $\mu$ F).

If  $R_2$  is higher than recommended, the involved risk will be that the voltage at pin 22 may rise to above the 1.5 V threshold voltage and the device will consequently fail to turn OFF when the mute line is brought down. About the stand-by, the time constant to be assigned in order to obtain a virtually pop-free transition has to be slower than 2.5V/ms.



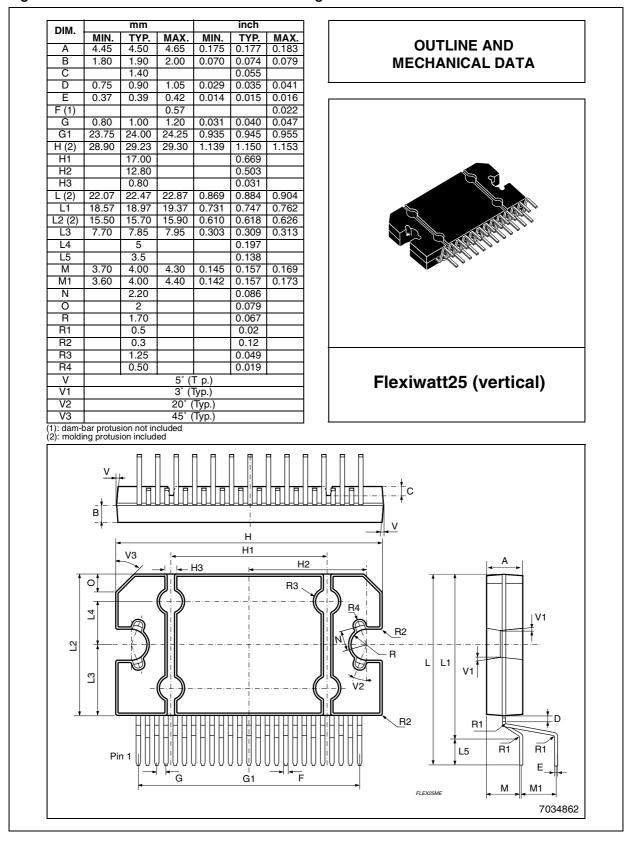


Figure 15. Flexiwatt 25 Mechanical Data & Package Dimensions

# **6 REVISION HISTORY**

## Table 5. Revision History

Date	Revision	Description of Changes
July 2005	1	First Issue



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